

## 65-volt, 7.5-amp, quad power half bridge

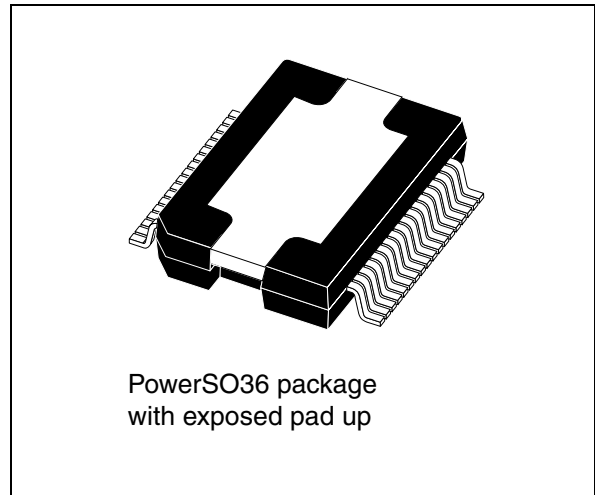
### Features

- Low input/output pulse-width distortion
- 200 mΩ  $R_{dsON}$  complementary DMOS output stage
- CMOS-compatible logic inputs
- Thermal protection
- Thermal warning output
- Undervoltage protection

### Description

STA516B is a monolithic quad half-bridge stage in Multipower BCD Technology. The device can be used as dual bridge or reconfigured, by connecting pin CONFIG to pins VDD, as a single bridge with double-current capability or as a half bridge (binary mode) with half-current capability.

The device is intended for the output stage of a stereo all-digital high-efficiency amplifier. It is capable of delivering 200 W + 200 W into 6-Ω loads with THD = 10% at  $V_{CC} = 51$  V or, in single BTL configuration, 400 W into a 3-Ω load with THD = 10% at  $V_{CC} = 52$  V.



The input pins have a threshold proportional to the voltage on pin VL.

The STA516B is aimed at audio amplifiers in Hi-Fi applications, such as home theatre systems, active speakers and docking stations.

It comes in a 36-pin PowerSO package with exposed pad up (EPU).

**Table 1. Device summary**

Order code	Temperature range	Package	Packaging
STA516B	0 to 90 °C	PowerSO36 EPU	Tube
STA516B13TR	0 to 90 °C	PowerSO36 EPU	Tape and reel

# 1 Introduction

The STA516B is a high performance quad half-bridge amplifier with the capability to drive up to 220 W <sup>(a)</sup> stereo into 3- to 8-ohm speakers from a single 50 V supply.

It offers the highest flexibility since it can be configured as a stereo-BTL, as a mono-BTL or as four channels of single-ended outputs to fit different application requirements.

It provides remarkably high levels of efficiency when driven by the FFX-patented 3-state pulse-width modulator embedded in STMs digital audio processors .

The device is self-protected by design. Overcurrent, overtemperature, under- and overvoltage protection are provided with an automatic recovery feature to safeguard the device and speakers against fault conditions that could damage the overall system.

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- a. The achievable output power depends on the thermal configuration of the final application.  
A high performance thermal interface material between the package exposed pad and the heat sink should be used in order to maximize output power levels

## 2 Pin description

Figure 1. Pin out

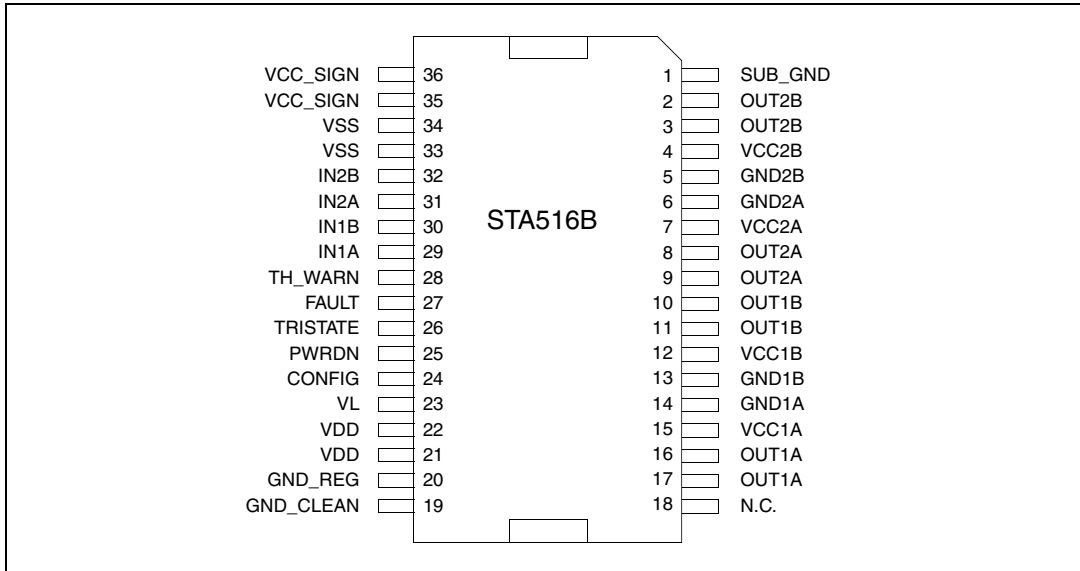


Table 2. Pin function

Pin	Name	Type	Description
1	GND_SUB	PWR	Substrate ground
2, 3	OUT2B	O	Output half bridge 2B
4	VCC2B	PWR	Positive supply
5	GND2B	PWR	Negative supply
6	GND2A	PWR	Negative supply
7	VCC2A	PWR	Positive supply
8, 9	OUT2A	O	Output half bridge 2A
10, 11	OUT1B	O	Output half bridge 1B
12	VCC1B	PWR	Positive supply
13	GND1B	PWR	Negative supply
14	GND1A	PWR	Negative supply
15	VCC1A	PWR	Positive supply
16, 17	OUT1A	O	Output half bridge 1A
18	N.C.	-	No internal connection
19	GND_CLEAN	PWR	Logical ground
20	GND_REG	PWR	Ground for regulator V <sub>DD</sub>
21, 22	VDD	PWR	5-V regulator referred to ground
23	VL	PWR	High logical state setting voltage, V <sub>L</sub>

Table 2. Pin function (continued)

Pin	Name	Type	Description
24	CONFIG	I	Configuration pin: 0: normal operation 1: bridges in parallel (OUT1A = OUT1B, OUT2A = OUT2B (If IN1A = IN1B, IN2A = IN2B))
25	PWRDN	I	Standby pin: 0: low-power mode 1: normal operation
26	TRISTATE	I	Hi-Z pin: 0: all power amplifier outputs in high impedance state 1: normal operation
27	FAULT	O	Fault pin advisor (open-drain device, needs pull-up resistor): 0: fault detected (short circuit or thermal, for example) 1: normal operation
28	TH_WARN	O	Thermal warning advisor (open-drain device, needs pull-up resistor): 0: temperature of the IC >130 °C 1: normal operation
29	IN1A	I	Input of half bridge 1A
30	IN1B	I	Input of half bridge 1B
31	IN2A	I	Input of half bridge 2A
32	IN2B	I	Input of half bridge 2B
33, 34	VSS	PWR	5-V regulator referred to +V <sub>CC</sub>
35, 36	VCC_SIGN	PWR	Signal positive supply

### 3 Electrical specifications

**Table 3. Absolute maximum ratings**

Symbol	Parameter	Value	Unit
$V_{CC\_MAX}$	DC supply voltage (pins 4, 7, 12, 15)	65	V
$V_{max}$	Maximum voltage on pins 23 to 32	5.5	V
$T_{j\_MAX}$	Operating junction temperature	0 to 150	°C
$T_{stg}$	Storage temperature	-40 to 150	°C

**Warning:** Stresses beyond those listed under “Absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “Recommended operating condition” are not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability. In the real application, power supplies with nominal values rated within the recommended operating conditions, may experience some rising beyond the maximum operating conditions for a short time when no or very low current is being drawn (amplifier in mute state, for instance). In this case the reliability of the device is guaranteed, provided that the absolute maximum rating is not exceeded.

**Table 4. Thermal data**

Symbol	Parameter	Min	Typ	Max	Unit
$T_{j\_case}$	Thermal resistance junction to case (thermal pad)	-	1	2.5	°C/W
$T_{warn}$	Thermal warning temperature	-	130	-	°C
$T_{jSD}$	Thermal shut-down junction temperature	-	150	-	°C
$t_{hSD}$	Thermal shut-down hysteresis	-	25	-	°C

**Table 5. Recommended operating conditions**

Symbol	Parameter	Min	Typ	Max	Unit
$V_{CC}$	Supply voltage for pins PVCCA, PVCCB	10	-	58	V
$T_{amb}$	Ambient operating temperature	0	-	90	°C

Unless otherwise stated, the test conditions for [Table 6](#) below are  $V_L = 3.3\text{ V}$ ,  $V_{CC} = 50\text{ V}$  and  $T_{\text{amb}} = 25\text{ °C}$

**Table 6. Electrical characteristics**

Symbol	Parameter	Test conditions	Min	Typ	Max	Unit
$R_{\text{dsON}}$	Power P-channel/N-channel MOSFET $R_{\text{dsON}}$	$I_{\text{dd}} = 1\text{ A}$	-	200	240	m $\Omega$
$I_{\text{dss}}$	Power P-channel/N-channel leakage $I_{\text{dss}}$	-	-	-	50	$\mu\text{A}$
$g_{\text{N}}$	Power P-channel $R_{\text{dsON}}$ matching	$I_{\text{dd}} = 1\text{ A}$	95	-	-	%
$g_{\text{P}}$	Power N-channel $R_{\text{dsON}}$ matching	$I_{\text{dd}} = 1\text{ A}$	95	-	-	%
$Dt_{\text{s}}$	Low current dead time (static)	see <a href="#">Figure 2</a>	-	10	20	ns
$Dt_{\text{d}}$	High current dead time (dynamic)	$L = 22\text{ }\mu\text{H}$ , $C = 470\text{ nF}$ $R_L = 8\text{ }\Omega$ , $I_{\text{dd}} = 4.5\text{ A}$ see <a href="#">Figure 3</a>	-	-	50	ns
$t_{\text{d ON}}$	Turn-on delay time	Resistive load	-	-	100	ns
$t_{\text{d OFF}}$	Turn-off delay time	Resistive load	-	-	100	ns
$t_{\text{r}}$	Rise time	Resistive load see <a href="#">Figure 2</a>	-	-	25	ns
$t_{\text{f}}$	Fall time	Resistive load see <a href="#">Figure 2</a>	-	-	25	ns
$V_{\text{IN-High}}$	High level input voltage	-	-	-	$V_L / 2 + 300\text{ mV}$	V
$V_{\text{IN-Low}}$	Low level input voltage	-	$V_L / 2 - 300\text{ mV}$	-	-	V
$I_{\text{IN-H}}$	High level input current	$V_{\text{IN}} = V_L$	-	1	-	$\mu\text{A}$
$I_{\text{IN-L}}$	Low level input current	$V_{\text{IN}} = 0.3\text{ V}$	-	1	-	$\mu\text{A}$
$I_{\text{PWRDN-H}}$	High level PWRDN pin input current	$V_L = 3.3\text{ V}$	-	35	-	$\mu\text{A}$
$V_{\text{Low}}$	Low logical state voltage (pins PWRDN, TRISTATE) (see <a href="#">Table 7</a> )	$V_L = 3.3\text{ V}$	0.8	-	-	V
$V_{\text{High}}$	High logical state voltage (pins PWRDN, TRISTATE) (see <a href="#">Table 7</a> )	$V_L = 3.3\text{ V}$	-	-	1.7	V
$I_{\text{VCC-PWRDN}}$	Supply current from $V_{\text{CC}}$ in power down	$V_{\text{PWRDN}} = 0\text{ V}$	-	-	2.4	mA
$I_{\text{FAULT}}$	Output current on pins FAULT, TH_WARN with fault condition	$V_{\text{pin}} = 3.3\text{ V}$	-	1	-	mA
$I_{\text{VCC-HiZ}}$	Supply current from $V_{\text{CC}}$ in 3-state	$V_{\text{TRISTATE}} = 0\text{ V}$	-	22	-	mA

**Table 6. Electrical characteristics (continued)**

Symbol	Parameter	Test conditions	Min	Typ	Max	Unit
$I_{VCC}$	Supply current from $V_{CC}$ in operation, both channels switching)	Input pulse width = 50% duty, switching frequency = 384 kHz, no LC filters	-	70	-	mA
$I_{OCP}$	Overcurrent protection threshold $I_{sc}$ (short-circuit current limit) <sup>(1)</sup>	-	7.5	8.5	10	A
$V_{UVP}$	Undervoltage protection threshold	-	-	7	-	V
$V_{OVP}$	Overvoltage protection threshold	-	61	62.5	-	V
$t_{pw\_min}$	Output minimum pulse width	No load	50	-	110	ns

1. See application note AN1994

**Table 7. Threshold switching voltage variation with voltage on pin VL**

Voltage on pin VL, $V_L$	$V_{LOW}$ max	$V_{HIGH}$ min	Unit
2.7	1.05	1.65	V
3.3	1.4	1.95	V
5.0	2.2	2.8	V

**Table 8. Logic truth table**

Pin TRISTATE	Inputs as per <a href="#">Figure 3</a>		Transistors as per <a href="#">Figure 3</a>				Output mode
	INxA	INxB	Q1	Q2	Q3	Q4	
0	x	x	Off	Off	Off	Off	Hi Z
1	0	0	Off	Off	On	On	Dump
1	0	1	Off	On	On	Off	Negative
1	1	0	On	Off	Off	On	Positive
1	1	1	On	On	Off	Off	Not used

### 3.1 Test circuits

Figure 2. Test circuit

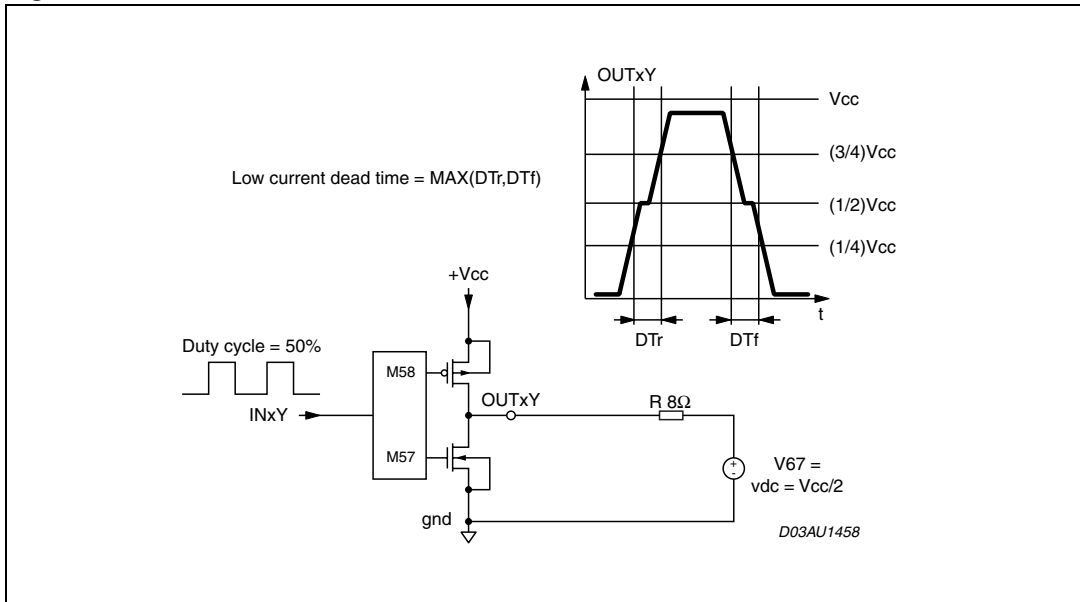
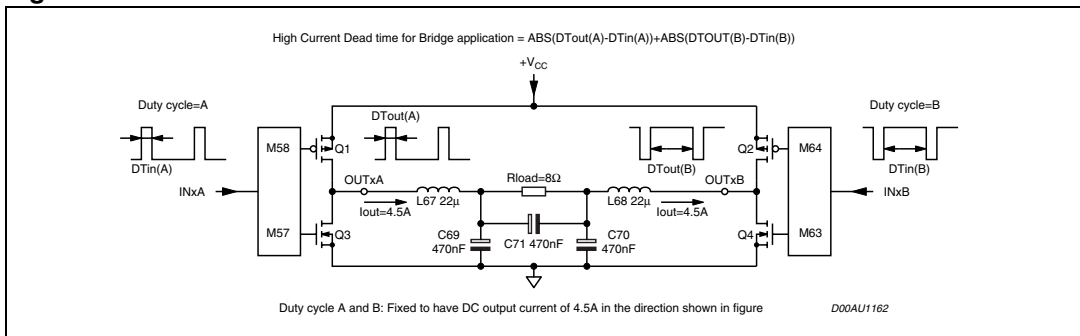


Figure 3. Current dead-time test circuit

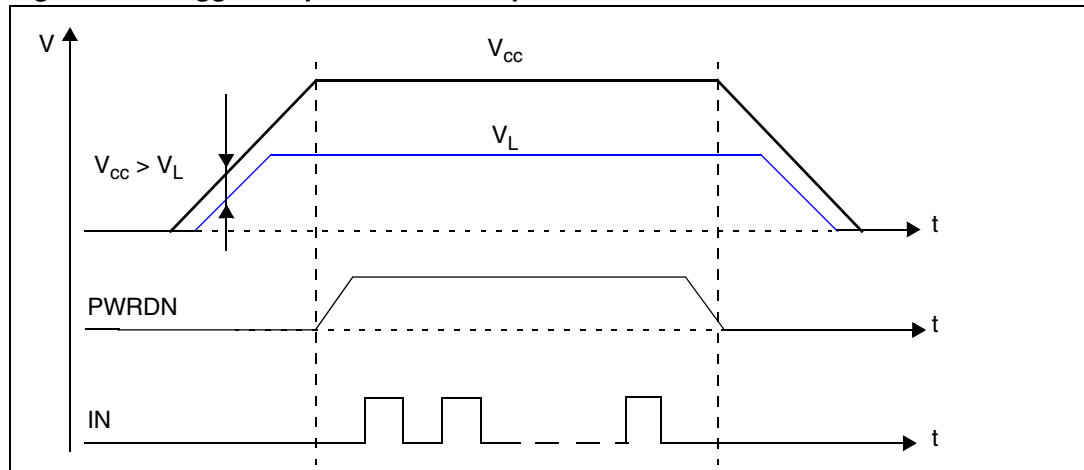




## 4 Power supply and control sequencing

To guarantee correct operation and reliability, the recommended power-on/off sequence as shown in [Figure 4](#) should be followed

**Figure 4. Suggested power-on/off sequence**



$V_{CC}$  should be turned on before  $V_L$ . This prevents uncontrolled current flowing through the internal protection diode connected between  $V_L$  (logic supply) and  $V_{CC}$  (high power supply), which could result in damage to the device.

PWRDN must be released after  $V_L$  is switched on. An input signal can then be sent to the power stage.

## 5 Applications information

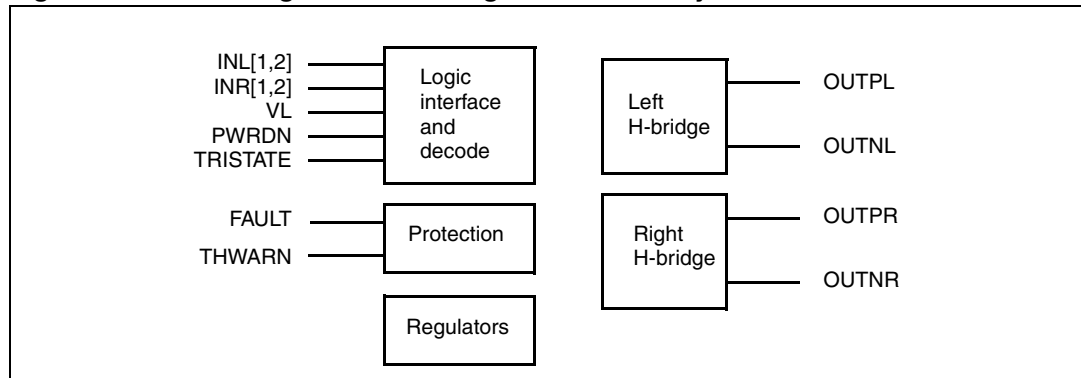
The STA516B is a dual channel H-bridge that is able to deliver 200 W per channel (into  $R_L = 6 \Omega$  with THD = 10% and  $V_{CC} = 51V$ ) of audio output power very efficiently. It operates in conjunction with a pulse-width modulator driver such as the STA321 or STA309A.

The STA516B converts ternary, phase-shift or binary-controlled PWM signals into audio power at the load. It includes a logic interface, integrated bridge drivers, high efficiency MOSFET outputs and thermal and short-circuit protection circuitry.

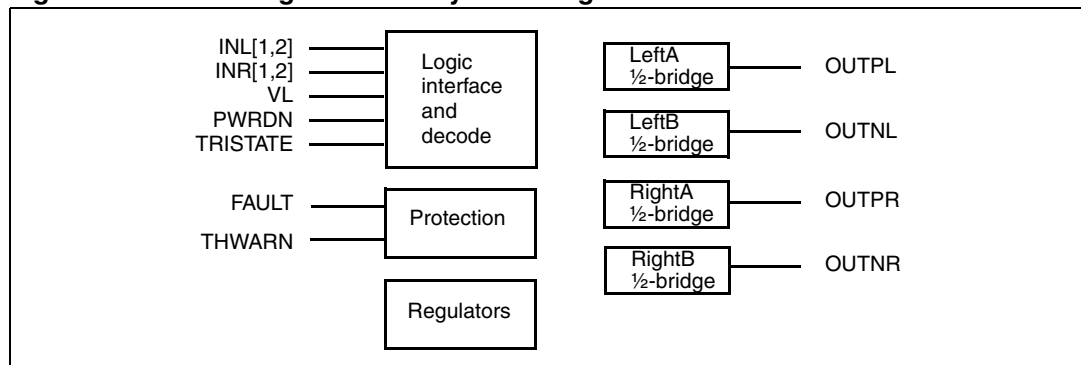
In differential mode (ternary, phase-shift or binary differential), two logic level signals per channel are used to control high-speed MOSFET switches to connect the speaker load to the input supply or to ground in a bridge configuration, according to the damped ternary modulation operation.

In binary mode, both full bridge and half bridge modes are supported. The STA516B includes overcurrent and thermal protection as well as an undervoltage lockout with automatic recovery. A thermal warning status is also provided.

**Figure 5. Block diagram of full-bridge FFX<sup>®</sup> or binary mode**



**Figure 6. Block diagram of binary half-bridge mode**



### 5.1 Logic interface and decode

The STA516B power outputs are controlled using one or two logic-level timing signals. In order to provide a proper logic interface, the VL input must operate at the same voltage as the FFX<sup>®</sup> control logic supply.

## 5.2 Protection circuitry

The STA516B includes protection circuitry for overcurrent and thermal overload conditions. A thermal warning pin (THWARN, pin 28, open drain MOSFET) is activated low when the IC temperature exceeds 130 °C, just in advance of thermal shutdown. When a fault condition is detected an internal fault signal immediately disables the output power MOSFETs, placing both H-bridges in a high-impedance state. At the same time the open-drain MOSFET of pin FAULT (pin 27) is switched on.

There are two possible modes subsequent to activating a fault.

- **Shutdown mode:** with pins FAULT (with pull-up resistor) and TRISTATE separate, an activated fault disables the device, signalling a low at pin FAULT output. The device may subsequently be reset to normal operation by toggling pin TRISTATE from high to low to high using an external logic signal.
- **Automatic recovery mode:** This is shown in the applications circuits below where pins FAULT and TRISTATE are connected together to a time-constant circuit (R59 and C58). An activated fault forces a reset on pin TRISTATE causing normal operation to resume following a delay determined by the time constant of the circuit. If the fault condition persists, the circuit operation repeats until the fault condition is cleared. An increase in the time constant of the circuit produces a longer recovery interval. Care must be taken in the overall system design not to exceed the protection thresholds under normal operation.

## 5.3 Power outputs

The STA516B power and output pins are duplicated to provide a low-impedance path for the device bridged outputs. All duplicate power, ground and output pins must be connected for proper operation.

The PWRDN or TRISTATE pin should be used to set all power MOSFETs to the high-impedance state during power-up until the logic power supply,  $V_L$ , has settled.

## 5.4 Parallel output / high current operation

When using the FFX<sup>®</sup> mode output, the STA516B outputs can be connected in parallel to increase the output current capability to the load. In this configuration the STA516B can provide up to 400 W into a 3- $\Omega$  load.

This mode of operation is enabled with pin CONFIG (pin 24) connected to pin VDD. The inputs are joined so that IN1A = IN1B, IN2A = IN2B and similarly the outputs OUT1A = OUT1B, OUT2A = OUT2B as shown in [Figure 8](#).

## 5.5 Output filtering

A passive 2nd-order filter is used on the STA516B power outputs to reconstruct the analog audio signal. System performance can be significantly affected by the output filter design and choice of passive components. Filter designs for 3- and 6- $\Omega$  loads are shown in the applications circuits of [Figure 7](#), [Figure 8](#) and [Figure 9](#).

## 5.6 Applications circuits

Figure 7. Typical stereo-BTL configuration for 200 W per channel

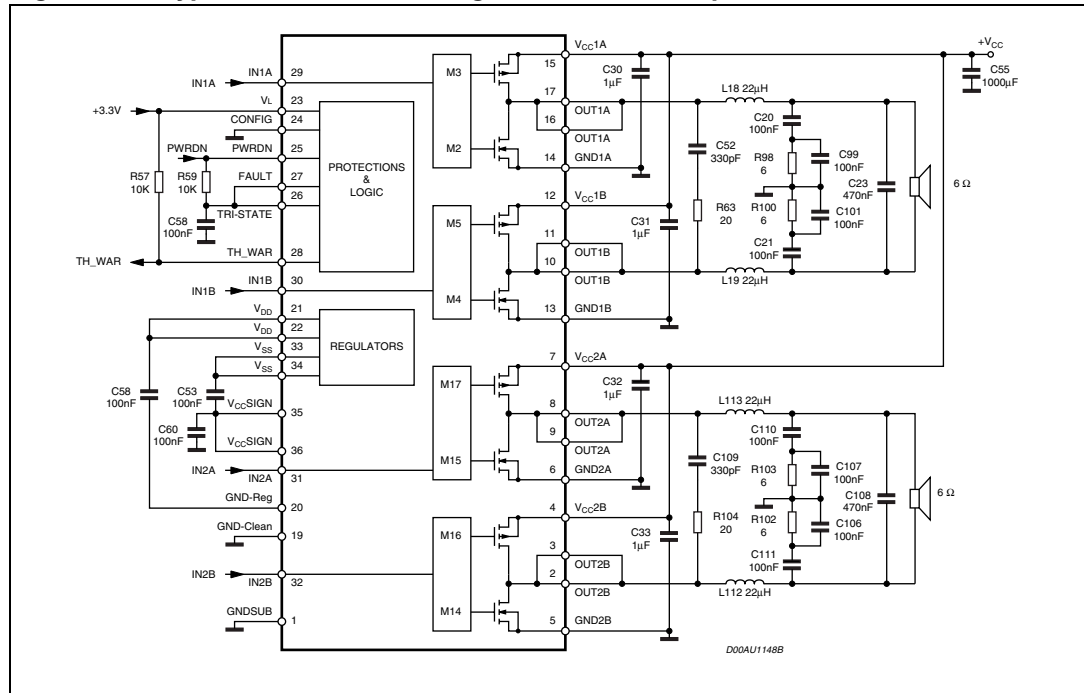


Figure 8 below shows a single-BTL configuration capable of giving 400 W into a 3-Ω load at 10% THD with  $V_{CC} = 52$  V. This result was obtained using the STA30X+STA50X demo board. Note that a PWM modulator as driver is required.

Figure 8. Typical single-BTL configuration for 400 W

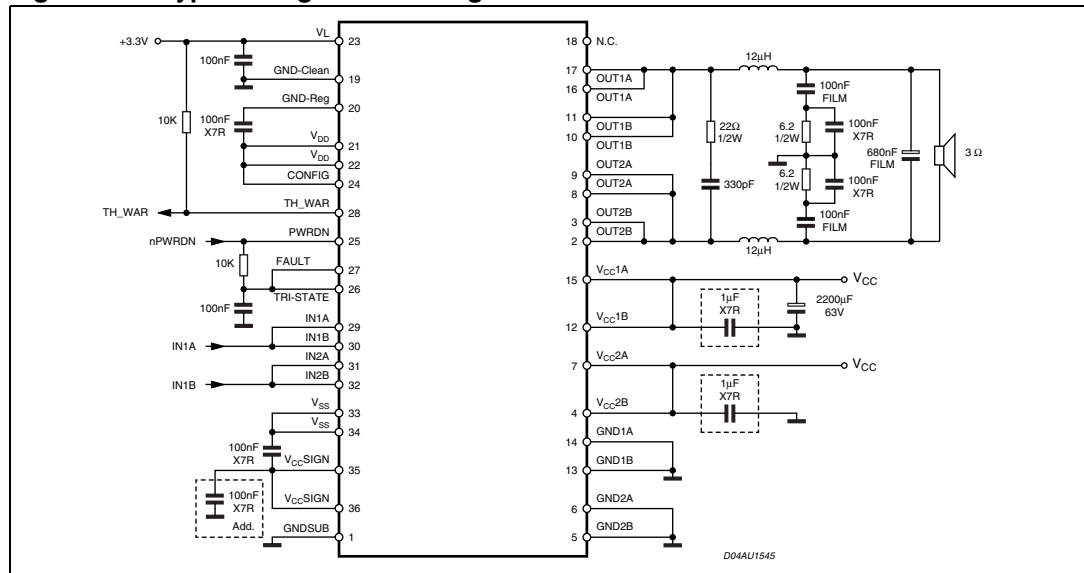
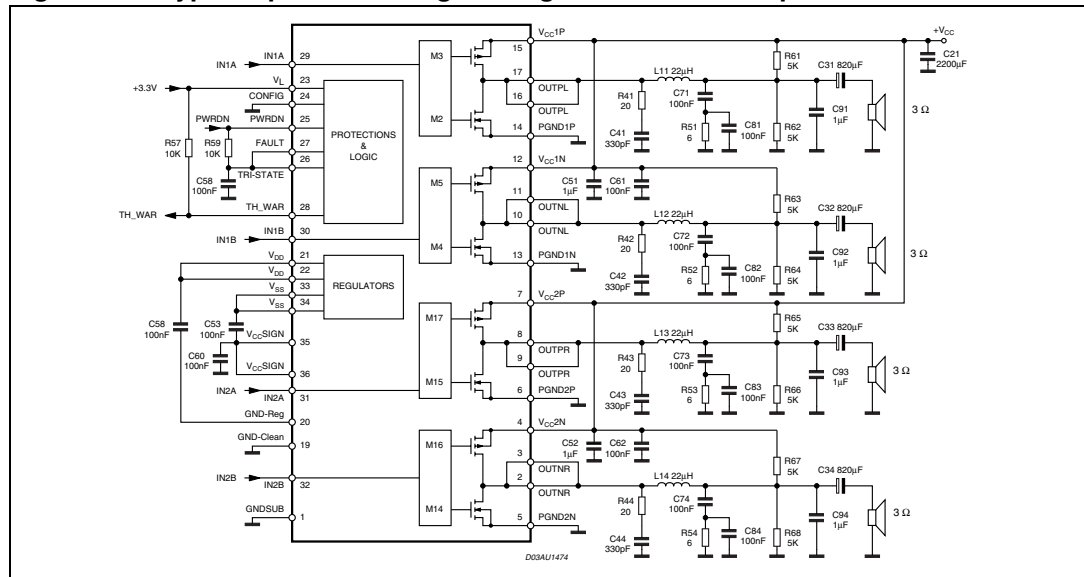


Figure 9. Typical quad half-bridge configuration for 100 W per channel



For more information, refer to the applications note AN1994.

# Package mechanical data

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Figure 10. PowerSO36 exposed pad up outline drawing

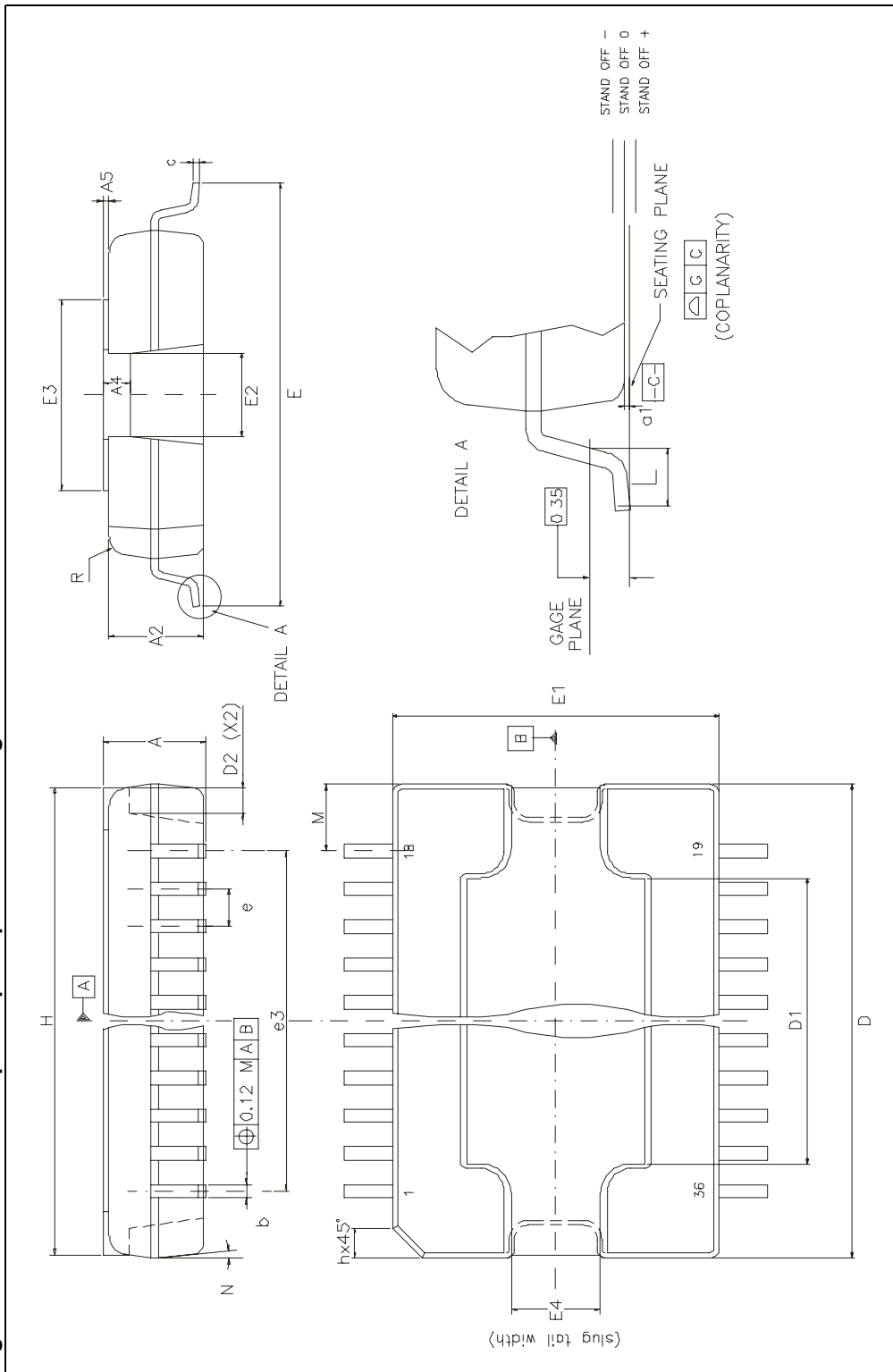


Table 9. PowerSO36 exposed pad up dimensions

Symbol	Dimensions in mm			Dimensions in inch		
	Min	Typ	Max	Min	Typ	Max
A	3.25	-	3.43	0.128	-	0.135
A2	3.10	-	3.20	0.122	-	0.126
A4	0.80	-	1.00	0.031	-	0.039
A5	-	0.20	-	-	0.008	-
a1	0.03	-	-0.04	0.001	-	-0.002
b	0.22	-	0.38	0.009	-	0.015
c	0.23	-	0.32	0.009	-	0.013
D	15.80	-	16.00	0.622	-	0.630
D1	9.40	-	9.80	0.370	-	0.386
D2	-	1.00	-	-	0.039	-
E	13.90	-	14.50	0.547	-	0.571
E1	10.90	-	11.10	0.429	-	0.437
E2	-	-	2.90	-	-	0.114
E3	5.80	-	6.20	0.228	-	0.244
E4	2.90	-	3.20	0.114	-	0.126
e	-	0.65	-	-	0.026	-
e3	-	11.05	-	-	0.435	-
G	0	-	0.08	0	-	0.003
H	15.50	-	15.90	0.610	-	0.626
h	-	-	1.10	-	-	0.043
L	0.80	-	1.10	0.031	-	0.043
M	2.25	-	2.60	0.089	-	0.102
N	-	-	10 degrees	-	-	10 degrees
R	-	0.6	-	-	0.024	-
s	-	-	8 degrees	-	-	8 degrees

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## 7 Revision history

**Table 10. Document revision history**

Date	Revision	Changes
01-Feb-2007	1	Initial release.
19-Mar-2007	2	Update to reflect product maturity
11-Aug-2009	3	Updated section Description on cover page.
16-Nov-2010	4	Modified presentation Updated <a href="#">Chapter 3: Electrical specifications on page 5</a> Added <a href="#">Chapter 5: Applications information on page 10</a>



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