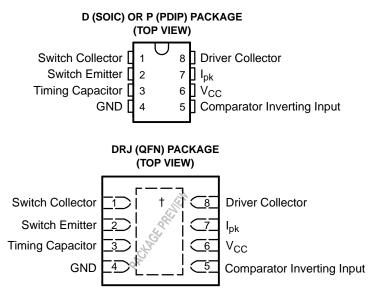


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## FEATURES

- Wide Input Voltage Range...3 V to 40 V
- High Output Switch Current...Up to 1.5 A
- Adjustable Output Voltage
- Oscillator Frequency...Up to 100 kHz
- Precision Internal Reference...2%
- Short-Circuit Current Limiting
- Low Standby Current



<sup>†</sup> Exposed thermal pad is connected internally to GND via die attach.

## **DESCRIPTION/ORDERING INFORMATION**

The MC33063A and MC34063A are easy-to-use ICs containing all the primary circuitry needed for building simple dc-dc converters. These devices primarily consist of an internal temperature-compensated reference, a comparator, an oscillator, a PWM controller with active current limiting, a driver, and a high-current output switch. Thus, the devices require minimal external components to build converters in the boost, buck, and inverting topologies.

The MC33063A is characterized for operation from  $-40^{\circ}$ C to  $85^{\circ}$ C, while the MC34063A is characterized for operation from  $0^{\circ}$ C to  $70^{\circ}$ C.

T <sub>A</sub>		PACKAGE <sup>(1)</sup>	ORDERABLE PART NUMBER	TOP-SIDE MARKING
	PDIP – P	Tube of 50	MC33063AP	MC33063AP
–40°C to 85°C	QFN – DRJ	Reel of 1000	MC33063ADRJR	PREVIEW
		Tube of 75	MC33063AD	M000004
	SOIC – D	Reel of 2500	MC33063ADR	— M33063A
	PDIP – P	Tube of 50	MC34063AP	MC34063AP
000 to 7000	QFN – DRJ	Reel of 1000	MC34063ADRJR	PREVIEW
0°C to 70°C	0010 0	Tube of 75	MC34063AD	140.40004
	SOIC – D	Reel of 2500	MC34063ADR	— M34063A

#### ORDERING INFORMATION

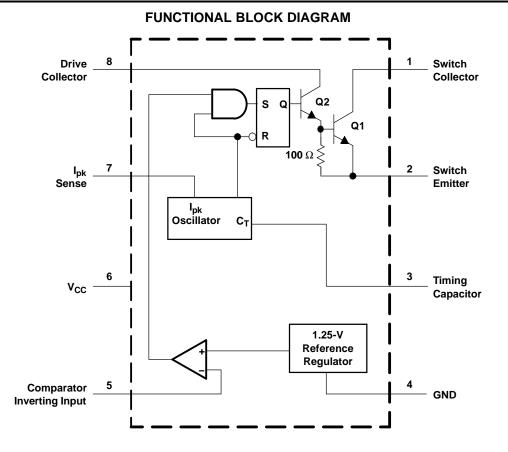
(1) Package drawings, standard packing quantities, thermal data, symbolization, and PCB design guidelines are available at www.ti.com/sc/package.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

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### Absolute Maximum Ratings<sup>(1)</sup>

over operating free-air temperature range (unless otherwise noted)

			MIN	MAX	UNIT
V <sub>CC</sub>	Supply voltage			40	V
V <sub>IR</sub>	Comparator Inverting Input voltage range		-0.3	40	V
V <sub>C(switch)</sub>	Switch Collector voltage			40	V
V <sub>E(switch)</sub>	Switch Emitter voltage	V <sub>PIN1</sub> = 40 V		40	V
V <sub>CE(switch)</sub>	Switch Collector to Switch Emitter voltage			40	V
V <sub>C(driver)</sub>	Driver Collector voltage			40	V
I <sub>C(driver)</sub>	Driver Collector current			100	mA
I <sub>SW</sub>	Switch current			1.5	А
		D package		97	
$\theta_{JA}$	Package thermal impedance <sup>(2)(3)</sup>	DRJ package		TBD	°C/W
		P package		85	
TJ	Operating virtual junction temperature			150	°C
T <sub>stg</sub>	Storage temperature range		-65	150	°C

(1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) Maximum power dissipation is a function of  $T_J(max)$ ,  $\theta_{JA}$ , and  $T_A$ . The maximum allowable power dissipation at any allowable ambient temperature is  $P_D = (T_J(max) - T_A)/\theta_{JA}$ . Operating at the absolute maximum  $T_J$  of 150°C can affect reliability.

(3) The package thermal impedance is calculated in accordance with JESD 51-7.

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### **Recommended Operating Conditions**

		MIN	MAX	UNIT
$V_{CC}$	Supply voltage	3	40	V
т	MC33063A	-40	85	ŝ
A	Operating free-air temperature MC34063A	0	70	

### **Electrical Characteristics**

 $V_{CC}$  = 5 V,  $T_A$  = full operating range (unless otherwise noted) (see block diagram)

### Oscillator

PARAMETER		TEST CONDITIONS	T <sub>A</sub>	MIN	TYP	MAX	UNIT
f <sub>osc</sub>	Oscillator frequency	$V_{PIN5} = 0 V, C_T = 1 nF$	25°C	24	33	42	kHz
I <sub>chg</sub>	Charge current	$V_{CC} = 5 V \text{ to } 40 V$	25°C	24	35	42	μΑ
I <sub>dischg</sub>	Discharge current	$V_{CC} = 5 V \text{ to } 40 V$	25°C	140	220	260	μΑ
I <sub>dischg</sub> /I <sub>chg</sub>	Discharge-to-charge current ratio	$V_{PIN7} = V_{CC}$	25°C	5.2	6.5	7.5	
V <sub>lpk</sub>	Current-limit sense voltage	$I_{dischg} = I_{chg}$	25°C	250	300	350	mV

### Output Switch<sup>(1)</sup>

PARAMETER		TEST CONDITIONS	T <sub>A</sub>	MIN	TYP	MAX	UNIT
V <sub>CE(sat)</sub>	Saturation voltage – Darlington connection	$I_{SW}$ = 1 A, pins 1 and 8 connected	Full range		1	1.3	V
V <sub>CE(sat)</sub>	Saturation voltage – non-Darlington connection <sup>(2)</sup>	$I_{SW}$ = 1 A, $R_{PIN8}$ = 82 $\Omega$ to $V_{CC},$ forced $\beta \sim 20$	Full range		0.45	0.7	V
h <sub>FE</sub>	DC current gain	I <sub>SW</sub> = 1 A, V <sub>CE</sub> = 5 V	25°C	50	75		
I <sub>C(off)</sub>	Collector off-state current	V <sub>CE</sub> = 40 V	Full range		0.01	100	μA

 Low duty-cycle pulse testing is used to maintain junction temperature as close to ambient temperature as possible.
 In the non-Darlington configuration, if the output switch is driven into hard saturation at low switch currents (<300 mA) and high driver</li> currents (≥30 mÅ), it may take up to 2 µs for the switch to come out of saturation. This condition effectively shortens the off time at frequencies >30 kHz, becoming magnified as temperature increases. The following output drive condition is recommended in the non-Darlington configuration:

Forced  $\beta$  of output switch = I<sub>C,SW</sub> / (I<sub>C,driver</sub> - 7 mA)  $\geq$  10, where ~7 mA is required by the 100- $\Omega$  resistor in the emitter of the driver to forward bias the V<sub>be</sub> of the switch.

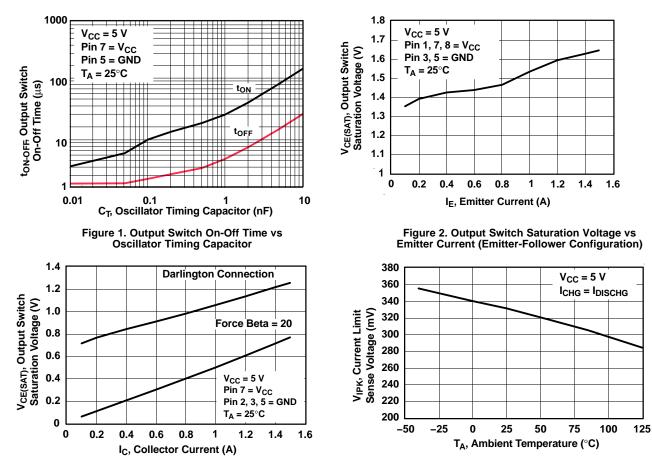
### Comparator

PARAMETER		TEST CONDITIONS	T <sub>A</sub>	MIN	TYP	MAX	UNIT
			25°C	1.225	1.25	1.275	V
V <sub>th</sub>	Threshold voltage		Full range	1.21		1.29	V
$\Delta V_{th}$	Threshold-voltage line regulation	$V_{CC} = 5 V$ to 40 V	Full range		1.4	5	mV
I <sub>IB</sub>	Input bias current	V <sub>IN</sub> = 0 V	Full range		-20	-400	nA

### **Total Device**

	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	MIN	MAX	UNIT
I <sub>CC</sub>	Supply current		Full range		4	mA

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**TYPICAL CHARACTERISTICS** 



Figure 4. Current-Limit Sense Voltage vs Temperature

Texas

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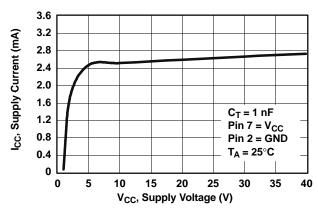


Figure 5. Standby Supply Current vs Supply Voltage

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## **TYPICAL CHARACTERISTICS (continued)**

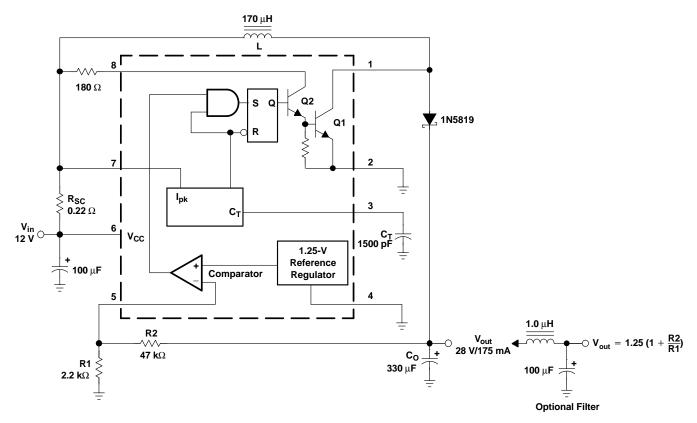
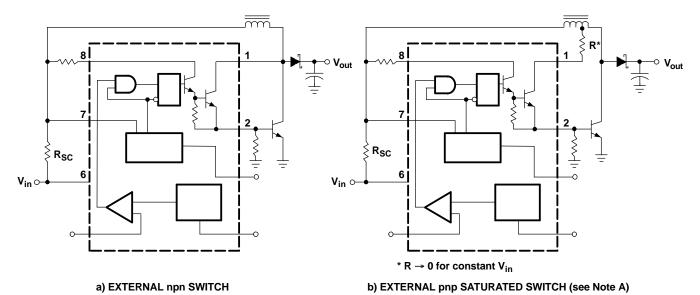


Figure 6. Step-Up Converter

TEST	CONDITIONS	RESULTS
Line regulation	$V_{IN} = 8$ V to 16 V, $I_O = 175$ mA	$30 \text{ mV} \pm 0.05\%$
Load regulation	$V_{IN} = 12 \text{ V}, I_{O} = 75 \text{ mA to } 175 \text{ mA}$	$10\mbox{ mV}\pm0.017\%$
Output ripple	V <sub>IN</sub> = 12 V, I <sub>O</sub> = 175 mA	400 mV <sub>PP</sub>
Efficiency	V <sub>IN</sub> = 12 V, I <sub>O</sub> = 175 mA	87.7%
Output ripple with optional filter	V <sub>IN</sub> = 12 V, I <sub>O</sub> = 175 mA	40 mV <sub>PP</sub>

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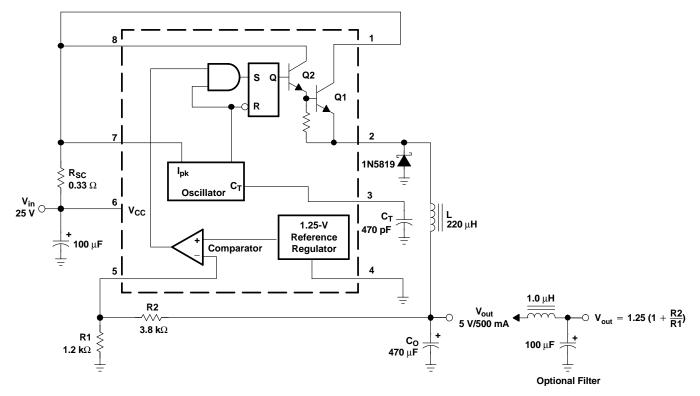
Texas

INSTRUMENTS www.ti.com

A. If the output switch is driven into hard saturation (non-Darlington configuration) at low switch currents (≤300 mA) and high driver currents (≥30 mA), it may take up to 2 µs to come out of saturation. This condition will shorten the off time at frequencies ≥30 kHz and is magnified at high temperatures. This condition does not occur with a Darlington configuration because the output switch cannot saturate. If a non-Darlington configuration is used, the output drive configuration in Figure 7b is recommended.

Figure 7. External Current-Boost Connections for I<sub>C</sub> Peak Greater Than 1.5 A

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#### Figure 8. Step-Down Converter

TEST	CONDITIONS	RESULTS
Line regulation	$V_{IN}$ = 15 V to 25 V, $I_O$ = 500 mA	$12 \text{ mV} \pm 0.12\%$
Load regulation	$V_{IN} = 25 \text{ V}, I_{O} = 50 \text{ mA to } 500 \text{ mA}$	3 mV ± 0.03%
Output ripple	V <sub>IN</sub> = 25 V, I <sub>O</sub> = 500 mA	120 mV <sub>PP</sub>
Short-circuit current	$V_{IN} = 25 \text{ V}, \text{ R}_{L} = 0.1 \Omega$	1.1 A
Efficiency	V <sub>IN</sub> = 25 V, I <sub>O</sub> = 500 mA	83.7%
Output ripple with optional filter	V <sub>IN</sub> = 25 V, I <sub>O</sub> = 500 mA	40 mV <sub>PP</sub>

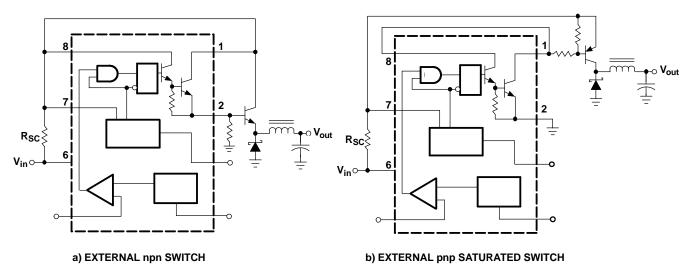
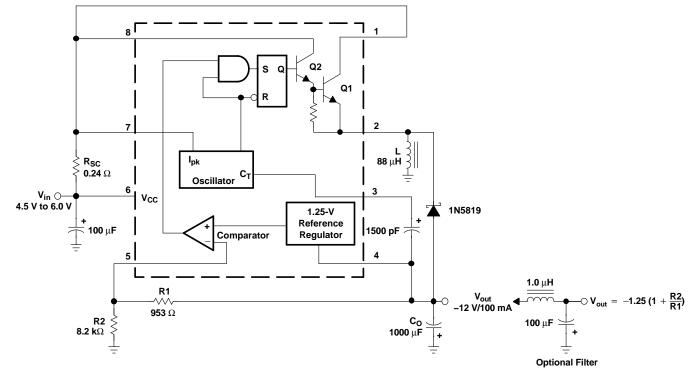


Figure 9. External Current-Boost Connections for  $I_C$  Peak Greater Than 1.5 A

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#### Figure 10. Voltage-Inverting Converter

TEST	CONDITIONS	RESULTS
Line regulation	$V_{IN} = 4.5 V$ to 6 V, $I_O = 100 mA$	$3 \text{ mV} \pm 0.12\%$
Load regulation	$V_{IN} = 5 V$ , $I_O = 10 mA$ to 100 mA	$0.022 \text{ V} \pm 0.09\%$
Output ripple	V <sub>IN</sub> = 5 V, I <sub>O</sub> = 100 mA	500 mV <sub>PP</sub>
Short-circuit current	$V_{IN} = 5 V, R_L = 0.1 \Omega$	910 mA
Efficiency	V <sub>IN</sub> = 5 V, I <sub>O</sub> = 100 mA	62.2%
Output ripple with optional filter	V <sub>IN</sub> = 5 V, I <sub>O</sub> = 100 mA	70 mV <sub>PP</sub>

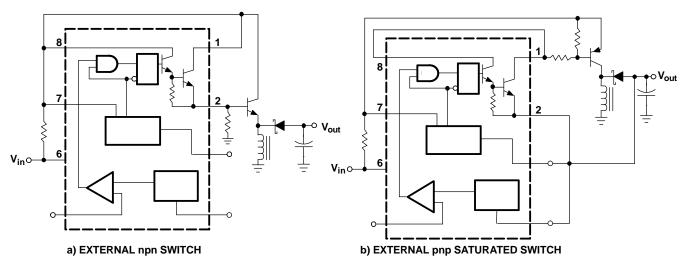


Figure 11. External Current-Boost Connections for I<sub>C</sub> Peak Greater Than 1.5 A

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### **APPLICATION INFORMATION**

CALCULATION	STEP UP	STEP DOWN	VOLTAGE INVERTING		
t <sub>on</sub> /t <sub>off</sub>	$\frac{V_{out}\ +\ V_{F}-V_{in(min)}}{V_{in(min)}\ -\ V_{sat}}$	$\frac{V_{out} + V_F}{V_{in(min)} - V_{sat} - V_{out}}$	$\frac{ V_{out}  \ + \ V_{F}}{V_{in} \ - \ V_{sat}}$		
$(t_{on} + t_{off})$	<u>1</u> f	<u>1</u> f	$\frac{1}{f}$		
t <sub>off</sub>	$\frac{t_{on} + t_{off}}{\frac{t_{on}}{t_{off}} + 1}$	$\frac{t_{on} + t_{off}}{\frac{t_{on}}{t_{off}} + 1}$	$\frac{t_{on} + t_{off}}{\frac{t_{on}}{t_{off}} + 1}$		
t <sub>on</sub>	$\begin{pmatrix} t_{on} \ + \ t_{off} \end{pmatrix} - \ t_{off}$	$\left( t_{on} \ + \ t_{off}  ight) - \ t_{off}$	$\left( t_{on} \ + \ t_{off}  ight) - \ t_{off}$		
CT	$4 imes 10^{-5}t_{on}$	$4 imes 10^{-5}t_{on}$	$4 imes10^{-5}t_{on}$		
I <sub>pk(switch)</sub>	$2I_{out(max)}\left(\frac{t_{on}}{t_{off}} + 1\right)$	2I <sub>out(max)</sub>	$2I_{out(max)}\left(\frac{t_{on}}{t_{off}} + 1\right)$		
R <sub>SC</sub>	0.3 I <sub>pk(switch)</sub>	0.3 I <sub>pk(switch)</sub>	0.3 I <sub>pk(switch)</sub>		
L <sub>(min)</sub>	$\left(\frac{\left(V_{in(min)}-V_{sat}\right)}{I_{pk(switch)}}\right)\!t_{on(max)}$	$\left(\frac{\left(V_{in(min)}-V_{sat}-V_{out}\right)}{I_{pk(switch)}}\right)\!t_{on(max)}$	$\left(\frac{\left(V_{in(min)} - V_{sat}\right)}{I_{pk(switch)}}\right)\! t_{on(max)}$		
Co	9 <mark>I<sub>out</sub>t<sub>on</sub> V<sub>ripple(pp)</sub></mark>	$\frac{I_{pk(switch)}(t_{on} + t_{off})}{8V_{ripple(pp)}}$	9 $rac{{\sf I}_{\sf out}{\sf t}_{\sf on}}{{\sf V}_{\sf ripple(pp)}}$		
V <sub>out</sub>	1.25(1 + <u>R2</u> ) See Figure 6	1.25(1 + <u>R2</u> ) See Figure 8	-1.25(1 + <u>R2</u> ) See Figure 10		

V<sub>sat</sub> = Saturation voltage of the output switch

V<sub>F</sub> = Forward voltage drop of the chosen output rectifier

The following power-supply parameters are set by the user:

V<sub>in</sub> = Nominal input voltage

Vout = Desired output voltage

I<sub>out</sub> = Desired output current

 $f_{min}$  = Minimum desired output switching frequency at the selected values of V<sub>in</sub> and I<sub>out</sub>

 $V_{ripple}$  = Desired peak-to-peak output ripple voltage. The ripple voltage directly affects the line and load regulation and, thus, must be considered. In practice, the actual capacitor value should be larger than the calculated value, to account for the capacitor's equivalent series resistance and board layout.

### MC33063A, MC34063A 1.5-A PEAK BOOST/BUCK/INVERTING SWITCHING REGULATORS SLLS636J-DECEMBER 2004-REVISED OCTOBER 2005



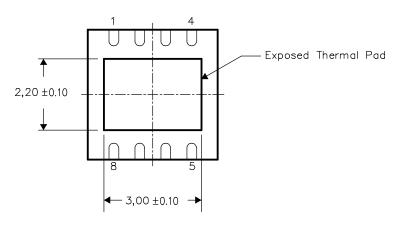
#### THERMAL PAD MECHANICAL DATA DRJ (S-PDSO-N8)

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB), the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to a ground plane or special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, Quad Flatpack No-Lead Logic Packages, Texas Instruments Literature No. SCBA017. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View

NOTE: All linear dimensions are in millimeters

Exposed Thermal Pad Dimensions

### **PACKAGING INFORMATION**

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Packag Qty	e Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
MC33063AD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
MC33063ADE4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
MC33063ADR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
MC33063ADRE4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
MC33063AP	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
MC33063APE4	ACTIVE	PDIP	Ρ	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
MC34063AD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
MC34063ADE4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
MC34063ADR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
MC34063ADRE4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
MC34063AP	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
MC34063APE4	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details. TBD: The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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## **MECHANICAL DATA**

MPDI001A - JANUARY 1995 - REVISED JUNE 1999



- NOTES: A. All linear dimensions are in inches (millimeters).
  - B. This drawing is subject to change without notice.
  - C. Falls within JEDEC MS-001

For the latest package information, go to http://www.ti.com/sc/docs/package/pkg\_info.htm



D (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in inches (millimeters).

B. This drawing is subject to change without notice.

C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).

D. Falls within JEDEC MS-012 variation AA.



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